# RNA trigger simulation 

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March 2, 2006

The aim of this work is to present the RNA software simulation and it's effect on the data analysis.

## 1 RNA trigger

The RNA trigger has been implemented in the trigger hardware since the 2001 data taking. Since run number 3865 RNA has been active in OR with the DNA trigger [M. Jabitzki -e-mail].

Its design followed the idea of the DNA trigger. In the latter the Ionisation Hodoscope was used to position the pions, in the RNA the Scintillator Fibre were used.

The input for the RNA trigger were SFD Vertical plane, Vertical Hodoscope Right and Left. The hit slabs from these detectors were send to a Neural Network box, which provides an output that is subsequently transformed in a trigger decision. All these informations may be found in : The Revised Neural Atomic trigger, 21/09/00 by S.Vlachos.

### 1.1 A little about the RNA

The most elaborate part in the RNA trigger simulation is the preparation of the hits to be sent to the NN algorithm RNA front-end. Thus we spend some time on this issue.

### 1.1.1 T0

In order to start the RNA front-end reading of the SFD information the T0 signal was used, like for the DNA. In contrast the T1 was used to latch the SFD signal in the TDCs. [A. Kulikov e-mail]. This may lead to differences when comparing trigger decision and SFD information (TDCs) in the offline.

### 1.1.2 SFD

All the information described here has been kindly given by M. Steinacher via e-mail and papers, see appendix.

The SFD vertical (X) has 240 column of fibres to be read out. Since the read-out cards could store only 64 bit, 4 cards (FDR CARDS) were used. FDR-Card number 1 contains the information about the first 64 column, the 2nd cards contains the information from the 65th column to the 128th, and so on. Later I will refer to "hit" as a column of the SFD that has been fired.

The number of hits in each of these FDR Cards are counted. In order to find the closest pair of hits additional C-Cards were installed after the FDR-Cards. In the first C-Card were stored the first 64 columns of the first FDR-card plus the first 16 columns of the second FDR card. The second C-Card was storing column 65 to $128+16$, in this
way we always have an overlap of two neighboring FDR cards to calculate the minimum distance between columns, without risking to loose information.

After this two layers of Cards we have for each Cards the following information :

- Number of hits.
- Minimum distance of two hits.

Particular cases are issued when there are:

- 1) no hits present in the card (Empty Flag)
- 2) more than 5 hits (Overflow Flag)
- 3) the minimal distance between hits is greater than 15 columns.


### 1.1.3 VH

For each VH detector (Left or Right) the slabs hit are collected. The information is kept if there are 1 or 2 slabs for each arm. Otherwise a special flag is set (overflow or empty).

### 1.2 Decision

The RNA decision was considered positive if :

- There was one overflow signal ( OR between the VH Left - VH Right and SFD ) That means more than 2 hits in VHR or VHL or more than 5 hits in SFD (sum of the 4 FDR Cards).
- The NN decision is positive. The input is : two SFD hits and one VHR and one VHL. In case there are more hits then all the combination are taken into account between the two SFD hits and the VH L and R.

The decision was considered negative if :

- There was an empty flag (Or of the SFD VHL and VHR).
- DEB15 Flag. This happens when the minimum distance found on ALL four C-Crads has a separation of 15 column at least.
- SHOM flag. Single Hit On Multiple C-Cards is activated if a single hit (distance 0 ) is detected in two or more C-Cards. If it is detected in only one Ccard, then the same hit is sent twice to the NN algorithm.


### 1.3 Possible error / malfunctioning

### 1.3.1 T0 and gate

The SFD is read out for the offline with a TDC signal while for the trigger the SFD is started by the ADC signal (TmTrigToADC). In Ariane, after the 2001 data taking the two signals coincide, but for the 2001 they are different. The literature recommends to use TmTrigToADC. As a test the TDC signals (TmTriggHodDC) were used, and not significant difference was found.

A gate of 40 ns is opened in order to read out the SFD columns, but the starting point of this 40 ns is tuned by hand and nowhere written, and could vary inside an interval of 60 ns . Thus this could introduce a different number of hits seen by the trigger as compared to the offline. The opening of the gate was fixed at hoc at 0 , and closed at 40 ns . Thus a hit is seen by the trigger if $0<($ TmTrigToADC - TimeHitScifi $)<40 \mathrm{~ns}$.

### 1.3.2 MASTER card

The logic of the trigger and all the possible cases are stored in the Master cards, no documents are available on how to read it again, and how exactly it was functioning.

### 1.3.3 The NN code

One version of the Neural Network code has been found by S. Vlachos (private communication) but he could not remember wether this was the one used for the training of the actual trigger in 2001.

## 2 Comparison with Data

- Data selection.

In order to evaluate the RNA simulation of the trigger, real data were selected when the RNA trigger was not in the decision of the trigger but the flag for it was in the hardware ( $\mathrm{T} 0+\mathrm{T} 4$ data in 2003). Then on this data the RNA trigger simulation was applied. The simulation takes as input the offline values of the variables, since the real hardware input is not stored in the data. The hardware decision was compared with the simulation decision and the number of coincidence counted. The events have to satisfy the Vertex Fit Tracking conditions using the DownStream and UpStream detectors.

- Comparison of hardware - software for RNA.

Of a total of 840198 events, 235059 events were reconstructed with the Vertex Fit Stream. Table 1 shows the result with the vertex Fit conditions. The RNA is apparently less well simulated than DNA. The main cause of the RNA simulation

|  | Software $=1$ Hardware $=0$ | Software $=0$ Hardware = 1 |
| :--- | :---: | :---: |
| DNA alone | $5.6 \%(13215)$ | $2.4 \%(5596)$ |
| RNA alone | $12.9 \%(30289)$ | $7.3 \%(17289)$ |

Table 1: Inefficiency of the DNA and RNA trigger simulations.
failure (in 19291 events out of 30289) is due to cases of too many hits seen by the simulation and not by the hardware trigger. In this particular case the software would give a positive decision while for the hardware it is negative. The cause could be multiple, an error in the choice of the gate for the aquisition of the hits in SFD, something missing in the Mother Card, or just a different limit in the overflow Flag, ie. more hits were allowed in the hardware.

- Comparison of hardware - software for DNA or RNA

Things are going much better for the OR of the simulations for DNA and RNA, that was actually used in the running.

| RNA OR DNA | Vertex tracking | Down Stream tracking |
| :--- | :---: | :---: |
| all $Q_{l}$ and all $Q_{x}$ and $Q_{y}$ | $93.4 \%$ | $93.4 \%$ |
| $Q_{l}<22 \mathrm{Mev}$ and $a b s\left(Q_{x}\right)<10 M e v$ | $99.7 \%$ | $99.6 \%$ |
| $Q_{l}<15 M e v$ and $a b s\left(Q_{x}\right)<4 M e v$ and $a b s\left(Q_{y}\right)<4 M e v$ | $99.7 \%$ | $99.5 \%$ |

Table 2: Efficiency of the DNA OR RNA trigger simulations.

For this test the Vertex tracking and the DownStream tracking were compared, see Table 2. The result are very similar.

### 2.1 Effect of RNA simulation on the MC data.

Sofar, in the analysis of the 2001 data only DNA was simulated in the MonteCarlos. Doing this we have lost some events and they show a $Q_{l}$ dependence, similar for $Q$, of less than $0.3 \%$ for $Q,\left|Q_{l}\right| \leq 15 \mathrm{MeV}$, figure 1 . This difference leads to a $0.5 \%$ more CC-background than with DNA simulation alone.

If we compare the MC events where the trigger simulations of T 0 and T 4 alone have been applied with the full trigger: $\mathrm{T} 0+\mathrm{T} 4+$ (RNA or DNA) we obtain figure 2 . We conclude that DNA.or.RNA introduce a slight $Q / Q_{l}$ dependance of less than $0.1 \%$ for $Q,\left|Q_{l}\right| \leq 15 \mathrm{MeV}$.



Figure 1: Coulumb Correlated MonteCarlo events. Ratio of events selected with the T0+T4+(DNA or RNA) / T0+T4+DNA as function of $Q_{l}$ and $Q$.


Figure 2: Coulumb Correlated MonteCarlo events. Ratio of events selected with the $\mathrm{T} 0+\mathrm{T} 4+\left(\mathrm{DNA}\right.$ or RNA) / T0+T4 as function of $Q_{l}$ and $Q$.

## 3 Documentation

- Jabitzki e-mail of the: Mon, 14 Nov 2005 18:22:53

Dear Colleagues,
After our last meeting the check of data-taking periods in 2001 with different triggers was performed. To find out active triggers content of trigger TDC3377 was checked with 2 different maskes:

- trigger itself
- trigger ! (ee—Lambda—Kaon—DNA(OR)—DNA1—DNA2) $=$ trigger !0x604a4

The last combination reveals whether some trigger in OR mode with DNA was active. The found periods were found to be consistent with 2001.run.logbook:

- T1pipicopl (T3—DNA) T4 ...-3477]
- T1pipicopl T4 [3480-3862]
- T1pipicopl (RNA—DNA) T4 [3865-...

Have a good day, Mikhail

- E-mail from Kulikov

Date: Fri, 18 Nov 2005 16:03:27 + 0300 (MSK) From: Kulikov ¡kulikov@nusun.jinr.rui To: Angela Benelli ¡benelli@mail.cern.ch¿ Subject: Re: Rna
Dear Angela, In order to latch the SFD signals in TDC we used T1. But to RNA system the SFD signals come independently. I suppose that for the RNA gating the same T0 was used like for DNA. The needed cabling for RNA was done not by me but by somebody else, most probably by Sotiris.
Regards, Anatoly

# DIRAC RNA-Trigger Hardware Overview 

Revision 1.3<br>Physics Basel, 14. December 2000, M. Steinacher

## 1. Introduction:

This paper gives an overview of the hardware needed for the Revised Neural Atomic trigger (RNA-trigger) at the DIRAC experiment (CERN / PS212). This trigger works similar to the already installed DIRAC Neural Atomic trigger (DNA-trigger); instead of the 16 channel Ionizing Hodoscope the 240 channel Vertical Scintillating Fibers (SciFi[1..240]) are used to locate the pionpair in front of the Spectrometer Magnet. After the Spectrometer Magnet the hit-maps of the two 18 channels Vertical Hodoscopes Left (VL[1..8]) and Right (VR[1..18]) are used to track the two pions. The analog detector signals are digitized with discriminators and preprocessed by the Preformat \& Decision Electronics (PDE) before the appropriate $N N$-patterns are transferred to the four $N N$-CARDs (built for the CP-LEAR experiment / CERN / PS195). The four $N N$-decisions are wired to the $P D E$ where the final $R N A$-decision is made. To reach a $R N A$-decision time of only several hundred nanoseconds, the whole PDE must be realized with fast digital electronics. The PDE as well as the $N N-C A R D$ s are located in the RNA-CRATE and they are loaded and controlled via the J1-VMEbus. Figure 1 shows the origin of the DIRAC detector-signals used for the RNA trigger.


For an introduction to the $R N A$-trigger see the paper 'The Revised Neural Atomic trigger', S.Vlachos, Basel, 21. September 2000.

The already installed DNA-trigger is documented in the following papers:

- 'The Dirac Neural Atomic trigger', S. Vlachos, Basel, 18. November 1998.
- 'Interface \& Decision CARD', M. Steinacher, Basel, 9. August 1999.
- 'The neural network first-level trigger for the DIRAC experiment', P. Kokkas et al., will be published in Nucl. Instr. and Meth. A

Figure 1 DIRAC detectors used for the RNA trigger: The analog signals from the Scintillating Fibers (SciFi[1..240]) and the Vertical Hodoscopes Left and Right (VL[1..18], VR[1..18]) are discriminated and the digital signals are connected to the RNA-trigger.

## 2. Layout:

The RNA-trigger hardware can be subdivided into two main parts: The Preformat \& Decision $\underline{E l e c t r o n i c s ~(P D E) ~ a n d ~ t h e ~ N e u r a l ~ N e t w o r k ~(N N) . ~ T h e ~ d e s c r i p t i o n ~ o f ~ t h e ~} N N$-electronics is not part of this article; it can be found in the following papers: 1 ) 'NN-CARD DESIGN', M. Steinacher, Basel, 11. September 1992. 2) 'NN-CARD PROTOTYPE', M. Steinacher, Basel, 27. September 1993.

The PDE converts each of the three detector hit-maps [SciFi(240), VL(18), VR(18)] into two binary numbers, representing the position of the hits. The two positions extracted from the 240 channel SciFi are coded in two 8 bit numbers [SciFi_POS1/2(16)] representing the two closest hits. The two positions of the 18 channel $V L$ are coded in two 5 bit numbers [VL_POS1(5), VL_POS2(5)] and the same scheme is also used for $V R$ [VR_POS1(5), VR_POS2(5)]. These six binary number are combined to build the four different $N N$-patterns. Each of the $N N$-patterns holds the two SciFi positions and a combination of one $V L$-and one $V R$-positions. This results in four different 26 bit wide $N N$-patterns. All $N N$-CARDs are loaded with the same weights, optimized to distinguish from a good and bad event. Further the PDE forms the final RNA-decision ( $R N A \_D E C$ ), manages the correct internal timing of the RNA-trigger sequence (SciFi_GATE, NN_CLK, V_CLK) and the communication with the 'DIRAC Trigger Control' via the following signals: T0_START, BUSY, VETO, RNA-DEC and STRB.


Figure 2
The Layout of the RNA hardware: The main block is the PDE part, which converts the detector hit-maps into four different $N N$-patterns. It also manages the correct internal timing as well as the communication with the 'DIRAC Trigger Control'. The NN is trained to be selective between good and bad events.

The PDE itself consists of total nine VME-cards (4 FDR-CARDs, 4 C-CARDs, 1 M-CARD) from three different types with the following functionality:

### 2.1 Fanout / Delay / Register Card (FDR-CARD):

The name $\underline{F a n o u t ~ / ~ D e l a y ~ / ~ R e g i s t e r ~ C a r d ~ a l r e a d y ~ i m p l i e s ~ i t s ~ f u n c t i o n a l i t y: ~ F i r s t ~ t h e ~ F D R-~}$ CARD receives/fanout the SciFi-signals arriving in differential ECL (dECL) level on the frontpanel. Since the SciFi-signals are not only used for the RNA-trigger, the fanout is necessary.

Then the received SciFi-signals are delayed by an adjustable amount of time (around $150 . .200 \mathrm{~ns}$ ). This is inevitable because the SciFi-signals are valid prior the lowest level trigger (TO_START) of the DIRAC experiment. The adjustable delay step-size can be around $5 . .10 \mathrm{~ns}$ because the SciFi_GATE, coming from the M-CARD, can be adjusted with a resolution of 0.5 ns (via the VMEbus) with respect to TO_START.

Finally the FDR-CARD registers the delayed SciFi-signals during the SciFi_GATE (dECL) is active. This is needed due to the fact that the SciFi-signals arrive in a time-window of about 40ns. The SciFi_GATE signal is generated on the M-CARD and it is connected to the FDRCARDs via the frontpanel. Since the same SciFi_GATE signal is used on all of the four FDRCARDs, it is necessary that the $110 \Omega$ termination resistors can be removed; the SciFi_GATE signal is daisy-chained from FDR-CARD \#4..\#1 and it is terminated on the last FDR-CARD \#1.

The entire 240 SciFi-signals are covered by four identical FDR-CARDs. One FDR-CARD is able to handle 64 SciFi inputs connected via the frontpanel. Since the FDR-CARD \#4 uses only 48 inputs out of the 64 inputs, the SciFi inputs which are left open should be defined low.

The TTL level outputs of the FDR-CARDs are connected via the J2-connector (96 pin, DIN41612) to the dedicated RNA-backplane where they are distributed to the C-CARDs. The outputs of the FDR-CARDs are able to drive termination on the C-CARDs of $180 \Omega$ to +5 V and $330 \Omega$ to $G N D$. Since the first 16 SciFi -signals are wired to two different $C$-CARDs these outputs are duplicated and buffered separately. Therefore 80 pins out of the 96 pins on the J2connector are used; the remainder are connected to GND.

### 2.3 Concentrator Card (C-CARD):

Each Concentrator Card (C-CARD) processes in parallel 80 registered TTL SciFi-signals from two adjacent FDR-CARDs. To get a continuos and constant detection performance over the whole width of all 240 SciFi channels, an overlap of 16 channels between two adjacent C-CARDs is implemented. The C-CARD \#1 processes the 64 registered TTL SciFi-signals from the FDR-CARD \#1 and the lowest 16 registered TTL SciFi-signals from FDR-CARD \#2. The same scheme is also used for the C-CARD \#2 and \#3; the C-CARD \#4 treats only the 48 registered TTL SciFi-signals from FDR-CARD \#4. Nevertheless all C-CARDs are completely identical.

The three main tasks of the $C-C A R D$ are:

- Find in the 80 bit hit-map the two hits which has the smallest distance (closest together). Send the detected minimum distance [MIN_DIST(4)] via dECL signals on the frontpanel to the $M-C A R D$. If a single hit is detected set the minimum distance to zero. Two adjacent hits correspond to a minimum distance of one. Up to a distance of 14 the algorithm is linear. The minimum distance of 15 indicates that the distance is equal or bigger than 15 or no hit is detected in the 80 bit hit-map.
- Count the number of hits in the lower 64 bit of the hit-map; this restriction makes sure that each hit is counted only once. Send the number of hits [HITS(4)] via dECL signals on the frontpanel to the $M$-CARD. If more than 15 hits are counted, it saturates at 15 .
- If the C-CARD is selected, the two absolute positions (each 8 bit) of the two closest hits are presented on the tri-state bus [SciFi-POS1/2(16)]. If there was only a single hit, two times the same position is presented. The absolute positions of the hits are calculate by using the VME address of the selected C-CARD. The select signals [SEL1..4, dECL] are generated on the M-CARD and they are connected to the CCARDs via the frontpanels. The outputs of the C-CARD are able to drive terminations on the $N N-C A R D$ s of $180 \Omega$ to +5 V and $330 \Omega$ to GND.

The distribution of the registered TTL SciFi-signals from the FDR-CARDs to the C-CARDs and the interconnection of the tri-state TTL bus [SciFi-POS1/2(16)] to the NN CARDs are realized on the RNA-backplane. All the 96 pins of the J2-connector are used for the signal interconnection [input: TTL SciFi-signals(80), output: SciFi-POS1/2(16)].

### 2.3 Master Card (M-CARD):

The Master Card (M-CARD) receives and processes the information from the four $C-C A R D$ s. All the signals from the C-CARDs are transmitted to the M-CARD in dECL format and are connected via the frontpanel. The M-CARD selects the C-CARD which has sent the smallest minimum distance [MIN_DIST1..4(4)]. This selection is accomplished by the outputs SEL1.. 4 (dECL) on the frontpanel of the M-CARD. If two or more C-CARDs send the same minimum distance the first $C$-CARD in the order \#1, \#2, \#3, \#4 is selected. Out of the four number of hits [HITS1..4(4)] the Total Number of Hits (TNH) in the SciFi detector can be calculated. If TNH is zero (no hit) the SciFi_EMPTY is activated; if TNH is bigger than 5 the SciFi_OVR is activated. A smallest minimum distance of 15 and the SciFi_EMPTY not activated, indicates one or more hit-pairs with a $\underline{\text { Distance } \underline{E} q u a l}$ or $\underline{B}$ igger than $\underline{15}$ (DEB15). By combining the four minimum distances [MIN_DIST1..4(4)] and the four number of hits [HITS1..4(4)], Single Hits $\underline{O n}$ Multiple (SHOM) C-CARDs can be detected. The four facts SciFi_EMPTY, SciFi_OVR, DEB15 and SHOM are needed for the final RNA-decision.

The $M$-CARD processes also the two detector hit-maps $V L(18)$ and $\operatorname{VR}(18)$. The $d E C L$ signals are connected to the $M-C A R D$ via the frontpanel; the $110 \Omega$ termination resistors can be removed and the signals can be daisy-chained to the already installed $D N A$-trigger. The $M$ $C A R D$ registers the $V L$ and $V R$ hit-maps on the rising edge of the $V_{-} C L K$ signal which is adjustable with a resolution of 0.5 ns (via VMEbus) with respect to TO_START. For each side of the Vertical Hodoscope, up to two hits can be processed. The M-CARD determines, separately for $V L$ and $V R$, the position of the hit(s) and they are coded in four 5 bit values [VL_POS1(5), VL_POS2(5), VR_POS1(5), VR_POS2(5)]. If only one hit is registered two times the same position is readout. Further the VL_EMPTY, VR_EMPTY (no hit) and VL_OVR, VR_OVR (more than 2 hits) are build to perform the final RNA-decision. The four positions of the Vertical Hodoscope are distributed to the $N N-C A R D s$ via the J2-connector and the RNA-backplane. Because each of the position is used on two different $N N$-CARDs they are duplicated and buffered separately. The outputs are able to drive terminations on the $N N$-CARDs of $180 \Omega$ to +5 V and $330 \Omega$ to GND. The NN_CLK signals [NN_CLK1..4] are buffered for each NN-CARD separately and are able to drive terminations of $82 \Omega$ to +5 V and $120 \Omega$ to GND. Therefore 44 pins out of the 96 pins on the J2-connector are used; the remainder are connected to GND.

The signal TO_START (dECL) is connected via the frontpanel and starts the programmable sequencer which steers the timing of the RNA-trigger sequence. An active VETO input signal ( $d E C L$ ) prevents the $R N A$-trigger from processing new events. The timing of the $R N A$-trigger is controlled by the signals Sci_Fi_GATE, V_CLK, NN_CLK and STRB. During the RNA-trigger is running the signal BUSY ( $d E C L$ ) on the frontpanel of the M-CARD is activated.

The four $N N$-decisions [ $N N \_D E C 1 . .4, d E C L$ ] are connected to the $M-C A R D$ via the frontpanel. The final $R N A$-decision is a $d E C L$ signal available on the frontpanel of the $M$ CARD and a 20ns STRB signal (dECL) indicates that the RNA-decision is valid. The final RNAdecision is carried out on the M-CARD using the following algorithm:

```
EMPTY = SciFi_EMPTY or VL_EMPTY or VR_EMPTY
SPREAD = DEB15 or SHOM
OVR = SciFi_OVR or VL_OVR or VR_OVR
NN_OR = NN_DEC1 or NN_DEC2 or NN_DEC3 or NN_DEC4
```


## 3. Signal Interconnections:

Figure 3 shows the signal interconnection diagram of the RNA-trigger. The most connections between the 13 cards are realized on the dedicated RNA-backplane with single ended TTL signals. The $d E C L$ detector signals are connected via the frontpanels of the FDR-CARDs and M-CARD. For daisy-chain purpose all dECL $110 \Omega$ termination resistors can be removed.


Figure 3 Signal Interconnection Diagram of the RNA-trigger: On the left side the interconnections on the RNA-backplane are shown (single ended TTL signals), and on the right the interconnections via the frontpanels (dECL signals). For loading, testing and debugging all cards are connect via the J1-VMEbus to the commercial available VMEbus-CPU.

## 4. Timing Diagram:

Figure 4 shows the typical timing diagram of the main signals during a complete RNA-trigger sequence. All the delay times are first rough estimations. To have all the signals valid after the TO_START the SciFi_DATA shows the delayed SciFi TTL detector-signals on the FDR-CARDs. The RNA-decision (RNA_DEC) stays valid until the next event (TO_START) is processed. The timing parameters SciFi_DELAY, SciFi_GATE and V_DELAY are programmable via the VMEbus with a resolution of 0.5 ns in a range of 12.5 ns to 140 ns . The $V L$ and $V R$ detector signals need a minimum time-delay of 20ns with respect to the rising edge of the TO_START-signal. If a SciFi_DELAY of zero is programmed, the rising edge of the SciFi_GATE signal is delayed about 18 ns with respect to the rising edge of T0_START.


Figure 4 Timing Diagram of the RNA-trigger: The RNA-trigger sequence is started by the TO_START signal; the STRB signal, which marks a valid RNA-decision, terminates the sequence. During the RNA-trigger is running, the BUSY signal is activated. All the delay times are prospective and not exactly defined yet.

## 5. RNA-CRATE:

The complete electronics for the RNA-trigger is installed in the RNA-CRATE with the following dimensions: width $=482 \mathrm{~mm}$, height $=440 \mathrm{~mm}$ and depth $=650 \mathrm{~mm}$. As shown in Figure 5 the crate is subdivided into two parts: 1) The left part has 3 slots with a height of 6 U and the standard VME depth of 160 mm . In this part the commercial available VMEbus-CPU for loading, testing and debugging is installed. 2) The right part provides 16 slots with a height of 9 U and a depth of 220 mm . In this part the FDR-, $C$-, $M$ - and $N N-C A R D s$ are installed. Between the two parts of the crate the communication is enabled via the J1-VMEbus. The RNA-CRATE is a modification of the old L3-NN-CRATE (SP746d) which is no longer used for the L3 experiment.

Three main things need to be modified before the crate can be used for the RNA-trigger

- The -5.2 V power supply has a maximum current of 5A only, which is to small for the RNAtrigger; therefore a stronger -5.2 V power supply has to be installed. The +5 V power supply has a maximum current of 300A, which should be sufficient for the RNA-trigger.
- The frontpanel with the power on/off switch and the control LEDs must be changed, so that all supply voltages ( $+5 \mathrm{~V},-5.2 \mathrm{~V},+12 \mathrm{~V},-12 \mathrm{~V}$ ) are indicated.
- The dedicated RNA-backplane has to be installed instead of the L3 Distribution Backplane.


Figure 5
The RNA-CRATE: The 16 slots in the right part have a height of 9U and a depth of 220 mm . To reach short interconnections on the dedicated RNA-backplane, each FDR-CARD is followed by a C-CARD. All slots on the left and right part of the RNA-crate are interconnected via the J1-VMEbus. Two 6U slots and three 9U slots are spare.

## 6. Board Specifications:

All three new CARDs for the RNA-trigger ( $F D R$-, $C$ - and $M$-CARD) are 9U VME boards with the dimensions of $366.7 \mathrm{~mm} \times 220 \mathrm{~mm}$. The J1 connector (96 pin) is used for the standard VMEbus communication with the CPU, located in the left part of the RNA-CRATE. The J2 connector (96 pin) is application specific; it is used for single ended TTL data in/output. The pin allocation is different for each of the three board types and is not defined yet. Unused pins of the $J 2$ connector are wired to GND. The $J 3$ connector ( $2 \times 16 \mathrm{pin}=32 \mathrm{pin}$ ) is a high current connector for the supply voltages $+5 \mathrm{~V},-5.2 \mathrm{~V}$ and GND. The connector type is a DIN41612 / D / 0904-132-6921. The boards have a mechanical coding system which prevents the user from placing a wrong cards to the different slots of the RNA-CRATE and possibly destroying the system. A recess on the back of the PCBs at different positions for the different card types makes this feature possible (see Figure 6). The RNA-CRATE is equipped with corresponding bolts at the different locations; for example a $C$-CARD fits only in the slots foreseen for this card type. The NN-CARDs are coded by the missing J3 connector.


Figure $6 \quad 9 U$ VME board: A recess on the back of the PCB at different positions for the different card types makes sure that only the appropriate CARDs fits in the corresponding slots. The pin connections of the high current connector J3 are given at the bottom.

## 7. Main Tasks:

Several tasks has to be executed before the RNA-trigger can be successfully utilize in the DIRAC experiment. At the moment the following main tasks can be identified (no entitlement to completeness). The order of the tasks is more less in sequence of time.

1. FDR-CARD: The detailed design (specifications, schematic diagram, PCB-layout, frontpanel, $C P L D s$ ), the production and the test of six FDR-CARDs have to be performed.
2. C-CARD: The detailed design (specifications, schematic diagram, PCB-layout, frontpanel, CPLDs), the production and the test (software required) of six C-CARDs have to be performed.
3. M-CARD: The detailed design (specifications, schematic diagram, $P C B$-layout, frontpanel, CPLDs), the production and the test (software required) of two $M-C A R D$ have to be performed.
4. C-, M- and NN-CARDs LUTs: All the LUT-files for the $C$-, $M$ and $N N-C A R D$ save to be generated. The LUT-files for the $C$-, M-CARDs are needed for complete testing and debugging the $C$ - and $M$-CARDs. The LUT-files for the $N N$-CARDs are used for the final test of the RNA-trigger in the lab.
5. RNA-BACKPLANE: The detailed design (specifications, schematic diagram, $P C B$-layout), the production and the test of two RNA-BACKPLANEs have to be performed.
6. RNA-CRATE: The two L3-NN-CRATEs have to be modified to meet the requirements of the $R N A$-trigger (see section 5 of this paper).
7. RNA Extension Board: The design (specifications, schematic diagram, PCB-layout) and the production of four RNA Extension Boards have to be performed. The RNA Extension Boards make possible to test and debug the different $R N A$-boards ( $C-, M$ and $N N-C A R D s$ ), plugged into the slots of the RNA-CRATE.
8. Test-patterns: Reasonable test-patterns for the final test in the lab (see next task) have to be created from good and bad events.
9. Final test in the lab: Software for testing and debugging the complete $R N A$-trigger has to be written. The RNA-CRATEs equipped with all $R N A$-boards will be tested in the lab. The final test is carried out with a fast pattern generator loaded with the test-patterns generated in task 8.
10. Installation and commissioning: The RNA-CRATE with the all RNA-boards have to be installed and cabled to the DIRAC experiment. The test-software written in task 9 has to be adapted to the VMEbus CPU. The proper functioning of the RNA-trigger with real detector signals has to be demonstrated.

# Master-CARD (M-CARD) SP 827 The Design <br> (For DIRAC RNA-Trigger / CERN PS212) 

Revision 2.0

Basel, 15. June 2001, M. Steinacher

## 1. Introduction:

This paper describes the design of the Master-CARD (M-CARD) SP 827 of the Revised Neural Atomic trigger (RNA-trigger). As Figure 1 shows, the M-CARD is the heart of the RNA-trigger. It processes the information from the four Concentrator CARDs (C-CARDs), it selects the C-CARD which has found the smallest minimum distance and it controls the correct timing of the RNAtrigger sequence. Further the $M-C A R D$ is needed to store and preprocess the digital detector-signals of the two Vertical Hodoscopes (left $V L$, right $V R$ ) before they are sent to the four $N N$-CARDs (built for the CP-LEAR experiment / PS195). The four NN-DECISIONs are wired to the M-CARD where the final RNA-DECISION is made. The M-CARD is realized with fast electronics to reach a RNAtrigger decision-time below 300 ns . The $M-C A R D$ is loaded, controlled and tested via the VMEbus. For an overview of the complete hardware needed for the RNA-trigger see 'DIRAC RNA-Trigger; Hardware Overview / M. Steinacher / Basel, 14. December 2000'.


Figure 1 The layout of the complete RNA hardware: Total 13 triple height VME-cards are needed. The $M-C A R D$ receives the information from the C-CARDs and the $N N-C A R D s$ and it interacts with the 'DIRAC Trigger Control'.

## 2. Topographical description:

The printed circuit board ( $P C B$ ) of the $M-C A R D$ is a eight layer board and has the same dimensions as the $N N-C A R D s$. The height of 366.7 mm corresponds to FASTBUS (triple height $V M E)$ and the depth of 220 mm is an EURO-standard (see Figure 2).


Figure 2
Topography of the M-CARD: The detector-signals, the C-CARD in/output-signals, the RNA control in/output-signals and the $N N$-decisions are connected via the frontpanel. On the rear side the J1-connector is used for the VMEbus and the $+12 \mathrm{~V} / 10 \mathrm{~mA}$ supply, the J2 connector for the NN-PATTERNs and NN-CLOCKs, and the J3 connector for the supply voltages (+5V/4.5A; -5.2V/2.5A; 0V).

### 2.1 Frontpanel description:

All the input-signals on the frontpanel of the M-CARD accept differential ECL (dECL) levels are normally terminated with $110 \Omega$. For daisy-chain options the $110 \Omega$ terminations ( $56 \Omega$ networks) can be removed and the signals have to be terminated with $110 \Omega$ at the end of the chain. All the output-signals on the frontpanel provide also dECL-levels and are able to drive $110 \Omega$ lines (twisted pair flat cable). For wired-OR options the pull-down resistors ( $270 \Omega$ networks) of the $d E C L$ output drivers can be removed.

The 18 detector-signals from the Vertical Hodoscope Left (VL1..18) are distributed to a standard 34pin flat cable connector (VL1..16) and to a standard 10pin flat cable connector (VL17..18) which holds also VR17.. 18 from the Vertical Hodoscope Right. The same scheme is also used for the 18 detector-signals from the Vertical Hodoscope Right (VR1..18).

The four NN-DECISIONs (NN1..4) have to be connected to the M-CARD with four short twisted pair cables ( $110 \Omega$ ) to the four 2pin-connectors. Also the two RNA control-inputs START and VETO are connected with 2pin-connectors. If the VETO is not used it can be left unconnected and the START is enabled. The activation of the START signal is indicated with a yellow $L E D$ and the BUSY-signal with a red $L E D$.

The eight RNA control-outputs RNA_DECISION, STROBE, SciFi_GATE, BUSY, SEL4, SEL3, SEL2 and SEL1 have to be connected to the M-CARD via short twisted pair cables $(110 \Omega)$ with 2pin-connectors.

The VME Card Number can be set by a Hex Switch in a range from Oh to Fh. The VMEbus communication is indicated with the green LED DTACK (data transfer acknowledge) and the red LED BERR (bus error).

The different modes of the $M-C A R D$ are indicated with the following LEDs:

- LOADED green: This LED shows that the look up tables (LUTs) and the registers are loaded and the M-CARD is now ready to switch to the WORK mode.
- +5 V OK green: This $L E D$ is activated if the +5 V supply is present and the RESET is not active.
- TEST yellow: This LED indicates that the M-CARD is in one of the six TEST modes. Different tests can be carried out via the VMEbus.
- LOAD red: This LED indicates that the M-CARD is prepared for loading the LUTs via the VMEbus.
- WORK green: This LED indicates that the M-CARD is working. This mode can only be reached after the LOADED is set. In this mode the M-CARD is ready to process events on the rising edge of the START-signal.
- S.S. yellow: This LED shows that the M-CARD is in the SINGLE SHOT mode. In this mode single events under real-time conditions can be analyzed via the VMEbus.


### 2.2 Rear description:

The rear of the M-CARD holds the two standard 96pin connectors $J 1, J 2$ and one 32pin J3 power-connector (DIN 41612/D/0904-132-6921). The connector J1 (upper) is used for the VMEbus and the $+12 \mathrm{~V} / 10 \mathrm{~mA}$ supply. The connector J2 (middle) holds the NN-PATTERNs (40pins), the NN -CLOCKs (4pins) and several GND connections (52pins). The connector J3 (lower) holds the power supply voltages $+5 \mathrm{~V} / 4.5 \mathrm{~A}$ (8pins), $-5.2 \mathrm{~V} / 2.5 \mathrm{~A}$ (8pins) and GND (16pins).

## 3. Functional Description:

Figure 3 shows the blockdiagram of the $M-C A R D$; for clarity also the four $N N-C A R D$ s and the selected C-CARD (one of the total four C-CARDs) are also shown. The digital detector-signals of the two Vertical Hodoscopes (VL1..18, VR1..18) are converted into four positions and they are combined with the two SciFi positions to the four different $N N$-PATTERNs.


Figure 3 Blockdiagram of the M-CARD: In the Pattern-Unit the four V-positions are generated from the detector-signals. The four minimum distances and the four number of hits (send by the four C-CARDs) are processed in the C-CARD-Unit. These values are used to select the CCARD with the smallest minimum distance and to generate empty, overflow and single hit information.

Each NN-PATTERN (26bit) holds information about two possible tracks in the two arms of the detector: Two SciFi-positions (each 8bit), one VL-position (5bit) and one VR-position (5bit). Each of the four off-line trained $N N-C A R D s$ searches for a good event in its pair of tracks.

The conversion of the detector hit-map to the $V$-positions is realized in the Pattern-Unit (PU) with two fast look up tables ( $V L L U T, V R L U T$ ). The $P U$ also detects if a hit-pattern has no hit (VL_EMPTY, VR_EMPTY) or more than two hits (VL_OVR, VR_OVR). These four flags are connected to the Decision-Unit.

The two SciFi-positions are directly transmitted (via a three-state bus) from the selected CCARD to the four NN-CARDs. The C-CARDs-Unit (CU) receives the four minimum distances (MIN_DIST1..4[0..3]) and the four numbers of hits (HITS1..4[0..3]) from the four C-CARDs. It selects (SEL1..4) the C-CARD which has send the smallest minimum distance and it generates several flag signals used in the in the Decision-Unit.

All the flag signals, coming from the $P U$ and $C U$, are used for the final RNA-DECISION in the Decision-Unit (DU). Also the four NN-DECISIONs (NN_DEC1..4) are fed to the DU where they are combined with the $A V A I L A B L E$ flags ( $V L \_1 / 2 P \_A V A, V R \_1 / 2 P \_A V A$ ) send by the $P U$. A valid RNA-DECISION is indicated by the strobe signal (STRB).

The Timing-Unit (TU) controls the correct synchronization of the Vertical Hodoscope inputregister ( $V$ _CLK), the Scintillator Fiber gate signal (SciFi_GATE) and the NN-CLKs with respect to the START signal (see timing diagram in Figure 4). All detector-signals need a minimum time-delay of 20 ns with respect to the rising edge of the START-signal (see Figure 4).

The VMEbus INTERFACE is the junction between the VMEbus and the 16bit on-board-bus (16bit OBB). The 16bit OBB is connected to all units and is used for loading and testing purpose.


Figure 4 Timing Diagram of the RNA-trigger: All the dECL signals are available on the frontpanel and the internal TTL signals are shown for clearness. The V_DELAY, the SciFi_DELAY and the SciFi_GATE are programmable with a fine resolution of 0.5ns.

### 3.1 Timing-Unit (TU):

The Timing-Unit (TU) controls the complete RNA-trigger sequence which is started by the rising edge of the START-signal and the STRB-signal indicates the end of the sequence. During an event is processed the BUSY-signal is activated and any further START-signals are ignored during this time. Figure 5 shows the block diagram of the $T U$.


Figure 5 The block diagram of the Timing-Unit: For fastest response all critical paths are realized with ECL logic. Several digital delays are used to adjust the proper timing of the signals.

The $T U$ generates the synchronized signals for the $V$ input-registers ( $V \_C L K$ ), for the Scintillator Fiber registers (SciFi_GATE) and for the NN-CARDs (NN_CLK). The two inputregister delays (V-DELAY, SciFi-DELAY) and the SciFi_GATE PW are VMEbus-programmable from $12.5 \mathrm{~ns} . .140 \mathrm{~ns}$ with a resolution of 0.5 ns ( 8 bit ). The two delays are started by the rising edge of the BUSY-signal and the SciFi_GATE PW by the rising edge of the SciFi-DELAY signal. At the minimum delay ( 12.5 ns ) all the corresponding detector-signals have to be valid at the latest 20 ns after the rising edge of the START-signal. At the maximum DELAY (140ns) all the corresponding detector-signals have to be valid at the latest 147.5 ns after the rising edge of the START-signal.

The time-delay from the rising edge of $V \_C L K$ or the falling edge of SciFi_GATE (the one which is the latest) to the $N N-C L K$ is on-board programmable from $6 \mathrm{~ns} . .120 \mathrm{~ns}$ ( $N N-D E L A Y$ ) via a jumper with a resolution of 6ns. A longer delay can be reached by changing the corresponding delay-chips. The $N N-D E L A Y$ delay is adjusted during test procedures and should not be changed during normal operation.

The STRB-signal is released after a certain time delay (STRB-DELAY) which is started by the rising edge of the $N N-C L K$. The STRB-DELAY is programmable via on board HEXswitches from 30ns..285ns with a resolution of 1 ns (8bit). This STRB-DELAY is adjusted during test procedures and should not be changed during normal operation. The width of the STRB-signal is set by a capacitor-trimmer to 20 ns and should not be changed.

The $T U$ also provides a BUSY-signal which indicates that the RNA-trigger is processing an event. The state of the BUSY-signal is shown with a red $L E D$ on the frontpanel. The $T U$ also accepts a VETO-signal which disables the RNA-trigger from accepting new events (the START signal is suppressed).

To prevent the $T U$ from blocking due to spikes or power supply glitches a Watchdog is installed. If the BUSY-signal is longer than 10 ms activated a reset is signal is released via the NEXT-signal.

In the WORK mode the TU is activated via the ENABLE-output and the test-outputs (T_START, T_NN_CLK) are inhibited.

In the SINGLE SHOT mode the normal RNA-trigger cycle is stopped after an event is completely processed. This is indicated by the READY-signal and a complete readout of the $M$ CARD and C-CARDs via the VMEbus can be carried out. A new event is only accepted after the NEXT/RESET-signal has been driven (high/low).

In the TEST_TU mode the TEST START-signal (T_START) and the TEST NN-CLK (T_NN_CLK) can be controlled from a test-register of the TU Logic \& Communication.

The high-speed part of the $T U$ is realized with ECL logic (Motorola 10k family) and FAST TTL logic (74F family). The control logic is implemented in a CPLD (Complex Programmable Logic Device). For further details see the CPLD documentation TU-CHIP.

### 3.2 Pattern-Unit (PU):

The Pattern-Unit (PU) converts the incoming detector-signal (VL1..18, VR1..18) from $d E C L$ to TTL and stores them on the rising edge of the $V_{-} C L K$, generated by the Timing-Unit (TU). To register the detector hit-patterns is necessary, because the detector-signals are typically valid for only 20 ns .

The $V L$ hit-pattern (VL1..18) is processed by a fast $256 \mathrm{k} \times 16 \mathrm{bit}$ SRAM working as look up table. The VL-LUT holds the two positions VL_1P (5bit) and VL_2P (5bit), the VL_EMPTY, the $V L \_O V R$, the $V L_{-} 1 P_{\_} A V A$ and the $V L \_2 P \_A V A$ information corresponding to the applied $V L$ hit-pattern. If no hit is registered in the $V L$ hit-pattern, the $V L_{-} E M P T Y$-bit is activated. If more than two hits are registered in the $V L$ hit-pattern, the $V L \_O V R$-bit is activated. If only one hit is registered in the $V L$ hit-pattern $V L \_1 P$ holds the position number, $V L \_2 P$ is set to zero ( 00000 b ) and only the $V L_{-} 1 P_{-} A V A$ flag is activated. If two hits are registered in the $V L$ hitpattern, $V L \_1 P$ holds the lower, $V L \_2 P$ the higher position number and both $V L \_1 P \_A V A$ and $V L \_2 P \_A V A$ flag are activated. If only one position is valid the non valid position is zero. All this information is needed by the Decision-Unit. The position VL_1P (5bit) is distributed to the $N N-C A R D 1$ and to the NN-CARD 2 via TTL Bus Drivers and the RNA Backplane via the J2 connector. The position VL_2P (5bit) is distributed to the NN-CARD 3 and to the NN-CARD 4.

The $V R$ hit-pattern (VR1..18) is processed in the same way as the $V L$ hit-pattern, only the distribution of the positions ( $V R \_1 P, V R \_2 P$ ) to the $N N-C A R D s$ is different: The position $V R \_1 P$ (5bit) is distributed to the $N N-C A R D 1$ and to the $N N$-CARD 3 and the position VR_2P (5bit) is distributed to the $N N-C A R D 2$ and to the $N N-C A R D 4$.

Before the $P U$ is able to work, the look up tables (VL-LUT, VR-LUT) have to be loaded with the precalculated values via the VMEbus. To speed-up the VL/VR-LUT loading process, first the LUTs can be completely filled with the OVERFLOW-bit set and afterwards only LUTaddresses with a different value than OVERFLOW can be written; the LUTs holds only around $1 \%$ of different values than OVERFLOW.

Figure 6 shows the block diagram of the Pattern-Unit. The high-speed part of the $P U$ is realized with ECL logic, fast SRAMs and FAST TTL (74F family) logic. The PU Logic \& Communication is implemented in a CPLD (Complex Programmable Logic Device). For further details see the CPLD documentation PU-CHIP.


Figure 6 Pattern-Unit (PU) block diagram: The detector hit-patterns are converted from dECL to TTL and are registered on the rising edge of the clock-signal (V_CLK). The precalculated relation between the hit-patterns (ADDRESS) and the two positions plus additional information (DATA) is stored in look up tables. This scheme leads to a very fast generation of the $N N$-patterns.

The $P U$ can be operated in different modes, which are selected via the $M$-Mode register (see also the part VMEbus INTERFACE):

- WORK: Detector hit-patterns are presented to the look up tables (ADDRESS) and its output-values (DATA) are sent to the $N N-C A R D s$ (via TTL Bus Drivers) and to the DU. This mode is selected during normal operation of the M-CARD. The input-registers (/V_REG_OE), the SRAMs (/V_RAM_OE) and the TTL Bus Drivers (/DRV_OE) are enabled.
- TEST_OVRALL: Test hit-patterns from test-registers of the PU Logic \& Communication are applied to the look up tables (ADDRESS) and its output-values (DATA) are sent to the $N N-C A R D s$ (via TTL Bus Drivers) and to the $D U$. This mode allows an internal overall test
of the RNA-trigger. The input-registers (/V_REG_OE) is disabled the SRAMs (/V_RAM_OE) and the TTL Bus Drivers (/DRV_OE) are both enabled.
- TEST_NN: Test output-values from test-registers of the PU Logic \& Communication are directly sent via the TTL Bus Drivers to the NN-CARDs. This mode allows testing the $N N$ CARDs. The input-registers ( $/ V_{-} R E G_{-} O E$ ) and the SRAMs ( $/ V_{-} R A M_{-} O E$ ) are both disabled, the TTL Bus Drivers (/DRV_OE) is enabled.
- TEST_PU: ADDRESSES from test-registers of the PU Logic \& Communication are applied to the look up tables and its output-values (DATA) are read via the PU Logic \& Communication. This mode allows the readout of the look up tables (SRAMs). The inputregisters (/V_REG_OE) and the TTL Bus Drivers (/DRV_OE) are both disabled, the SRAMs (/V_RAM_OE) is enabled.
- LOAD_SRAM: ADDRESSES and DATA are applied from test-registers of the PU Logic \& Communication to the SRAMs. The DATA is written by driving (low/high) the Write Enable lines (/VL_RAM_WE, /VR_RAM_WE). This mode allows the loading of the look up tables. The input-registers ( $V \_R E G \_O E$ ), the SRAMs (/V_RAM_OE) and the TTL Bus Drivers (/DRV_OE) are disabled, only the Write Enable lines are driven.

| Mode | $\overline{\text { V_REG_OE }}$ | $\overline{V_{-} R A M \_O E ~}$ | $\overline{\text { DRV_OE }}$ | $\overline{\text { VL/VR_RAM_WE }}$ |
| :---: | :---: | :---: | :---: | :---: |
| WORK | 0 | 0 | 0 | 1 |
| TEST_OVRALL | 1 | 0 | 0 | 1 |
| TEST_NN | 1 | 1 | 0 | 1 |
| TEST_PU | 1 | 0 | 1 | 1 |
| LOAD_SRAM | 1 | 1 | 1 |  |

Figure 7 Mode Truth Table for the PU: For the different modes the corresponding state of the PU control-signals are shown in this table.

### 3.3 C-CARDs-Unit (CU):

The C-CARDs-Unit (CU) receives the four minimum distances (MIN_DIST1..4[0..3]) and the four number of hits (HITS1..4[0..3]) from the four C-CARDs and converts the signals from $d E C L$ to TTL. These values must not be registered because the SciFi-hitmaps are stored on the FDR-CARDs and the data sent by the four $C$-CARDs are stable until a new event is processed.

The $C U$ selects (SEL1..4) the C-CARD which has send the smallest minimum distance and this card presents the two SciFi positions on the three-state bus on the RNA-backplane. If two or more $C-C A R D s$ send the same minimum distance, the lowest $C-C A R D$ is selected. A minimum distance of zero indicates a single hit on the corresponding C-CARD; a single hit is only preferred if the other $C$-CARDs send a minimum distance of 15 . This selection task is implemented by a fast 64 k x 4 bit SRAM working as look up table (MIND-LUT) and a multiplexer. The multiplexer makes sure that, under all conditions, only one C-CARD can be selected and therefore no bus conflict can occur. The select signals are active high $d E C L$ levels. If all minimum distances are 15 the signal MIN_DIST15 gets activated; if one or more minimum distance(s) are zero and the remainder are 15, the signal looks like a single hit (LL_SINGLE_HIT) is activated. These two flags are processed by the Decision-Unit.

Further the $C U$ evaluates the four number of hits (HITS1..4[0..3]), found in each C-CARD, using a fast $64 \mathrm{k} \times 4 \mathrm{bit}$ SRAM as look up table (HITS-LUT). If the sum of all the hits is zero, the signal SciFi_EMPTY gets activated, if it is exactly one the signal SINGLE_HIT is activated and if it is more than 5 the signal SciFi_OVR is set. These three flags are also evaluated by the Decision-Unit.

Before the $C U$ is able to work, the look up tables (MIND-LUT, HITS-LUT) have to be loaded with the precalculated values via the VMEbus. To speed-up the HITS-LUT loading process, first the LUT can be completely filled with the OVERFLOW-bit set and afterwards only LUT-addresses with a different value than OVERFLOW can be written.

Figure 8 shows the block diagram of the $C U$. The high-speed part of the $C U$ is realized fast SRAMs and FAST TTL (74F family) logic. The CU Logic \& Communication is implemented in a CPLD (Complex Programmable Logic Device). For further details see the CPLD documentation CU-CHIP.


Figure 8 C-CARDs-Unit (CU) block diagram: The information from the C-CARDs are processed with fast SRAMs used as look up tables. The multiplexer makes sure that only one C-CARD can be selected at one time and therefore no bus conflict can occur.

The $C U$ can be operated in different modes, which are selected via the $M$-mode register (see also the part VMEbus INTERFACE):

- WORK: The values from the C-CARDs are presented to the look up tables (ADDRESS) and its output-values (DATA) are used to control the multiplexer and the Decision-Unit. This mode is selected during normal operation of the $M$-CARD. The input-buffer (/HM_BUF_OE), the SRAMs (/HM_RAM_OE) and the Multiplexer (/MUX_E) are enabled.
- TEST_OVRALL: Test hit-patterns from test-registers of the CU Logic \& Communication are applied to the look up tables (ADDRESS) and its output-values (DATA) are sent to the multiplexer and the Decision-Unit. This mode allows an overall test of the RNA-trigger. The input-buffer ( $/ H M_{-} B U F_{-} O E$ ) is disabled and the SRAMs (/HM_RAM_OE) and the Multiplexer (/MUX_E) are both enabled.
- TEST_CU: ADDRESSES from test-registers of the CU Logic \& Communication are applied to the look up tables and its output-values (DATA) are read via the CU Logic \& Communication. This mode allows the readout of the look up tables (SRAMs). The inputbuffer (/HM_BUF_OE) and the Multiplexer (/MUX_E) are both disabled and the SRAMs (/HM_RAM_OE) is enabled.
- LOAD_SRAM: ADDRESSES and DATA are applied from test-registers of the CU Logic \& Communication to the SRAMs. The DATA is written by driving (low/high) the Write Enable lines (/HITS_RAM_WE, /MIND_RAM_WE). This mode allows the loading of the look up tables. The input-buffer (/HM_BUF_OE), the SRAMs (/HM_RAM_OE) and the Multiplexer (/MUX_E) are disabled, only the Write Enable lines are driven.

| Mode |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: |
| HM_BUF_OE | $\overline{\text { HM_RAM_OE }}$ | $\overline{\text { MUX_E }}$ | $\overline{\text { HM_RAM_WE }}$ |  |
| WORK | 0 | 0 | 0 | 1 |
| TEST_OVRALL | 1 | 0 | 0 | 1 |
| TEST_CU | 1 | 0 | 1 | 1 |
| LOAD_SRAM | 1 | 1 | 1 | $\square$ |

Figure 9 Mode Truth Table for the CU: For the different modes the corresponding state of the $C U$ control-signals are shown in this table.

### 3.4 Decision-Unit (DU):

The Decision-Unit ( $D U$ ) received the four $N N$-DECISIONs from the four $N N-C A R D s$ and converts the signals from $d E C L$ to $T T L$. These signals are directly fed to the $D U \operatorname{Logic} \&$ Communication. The $N N$-DECISIONs are quasi-static signals ( $N N$-Pattern is stable until a next event is processed) and therefore no register is needed. Figure 10 shows the block diagram of the $D U$.


Figure 10 Decision-Unit (DU) block diagram: In the DU the NN-DECISIONs and the flags of the Pattern-Unit and of the C-CARDs-Unit are combined to the final RNA-DECISON.

The $D U$ is realized with ECL logic, FAST TTL logic and a CPLD (Complex Programmable Logic Device). For further details see the CPLD documentation $D U-C H I P$.

The $D U$ can be operated in different modes, which are selected via the $M$-mode register (see also the VMEbus INTERFACE):

- WORK: The external NN-DECISIONs are processed and the RNA-DECISION output is driven by the $D U$ Logic. This mode is selected during normal operation of the M-CARD.
- TEST_OVRALL: Test patterns from test-registers of the DU Logic \& Communication are applied to the NN-DECISIONs inputs. The RNA-DECISION output is driven by the DU Logic. This mode allows an overall test of the RNA-trigger.
- TEST_DU: The input and output of the DU Logic are controlled via test-registers of the DU Logic \& Communication (see Figure 11). This mode allows complete testing the Decision Logic as well as the RNA-DECISION output driver.


Figure 11 The DU Logic \& Communication Chip in the TEST_DU mode: In this test mode the inputs of the LOGIC can be stimulated from test-registers and the outputs of the LOGIC can be read via a test-registers. The output-pins of the chip are also controlled from test-registers.

In the Empty \& Overflow Detection Logic the EMPTY and the OVR-signal are generated by an or-function of the three detector EMPTY-signals (VL_EMPTY or VR_EMPTY or SciFi_EMPTY) and an or-function of the three detector OVR-signals (VL_OVR or VR_OVR or SciFi_OVR).

In the $N N$-Decision Validation Logic the four NN_DECISIONs (NN_DEC1..4) are combined with the two corresponding AVAILABLE flags (VL_1P_AVA, VL_2P_AVA, $\left.V R \_1 P \_A V A, V R \_2 P \_A V A\right)$ from the Pattern-Unit. This ensures that only NN_DECISIONs with the corresponding positions available are taken into the final decision. The or-function of the validated $N N$-Decisions (NN_VAL1..4) results in the signal $N N \_O R$ :

```
NN_VAL1 = NN_DEC1 and VL_1P_AVA and VR_1P_AVA
NN_VAL2 = NN_DEC2 and VL_1P_AVA and VR_2P_AVA
NN_VAL3 = NN_DEC3 and VL_2P_AVA and VR_1P_AVA
NN_VAL4 = NN_DEC4 and VL_2P_AVA and VR_2P_AVA
NN_OR = NN_VAL1 or NN_VAL2 or NN_VAL3 or NN_VAL4
```

The DEB15 and SHOM Logic generates the following two signals: $\underline{\text { Distance } \underline{E q u a l} \text { or }}$ Bigger than $\underline{15}$ (DEB15), which indicates that the two closest hits found on all the four $C$ CARDs has a distance of 15 or more. The signal $\underline{S i n g l e} \underline{H} i t \underline{O} \underline{\text { Multiple C-CARDs (SHOM) }}$ identify the case of a single hit on two or more different C-CARDs. These two signals are generated by the following equations:

```
DEB15 = MIN_DIST15 and not SciFi_EMPTY
SHOM = LL_SINGLE_HIT and not SINGLE_HIT
```

The Final Decision Logic generates the RNA-DECISION out of the following signals: EMPTY, OVR, NN_OR, DEB15 and SHOM. The OVERFLOW-condition can be enabled or disabled via a bit in the control register of the DU Logic \& Communication.

If the OVERFLOW-condition is enabled the following logic is applied:
RNA_DEC = NN_OR and not (EMPTY or DEB15 or SHOM or OVR)

If the OVERFLOW-condition is disabled the following logic is applied:
RNA_DEC = OVR or [NN_OR and not (EMPTY or DEB15 or SHOM)]
An activated RNA_DECISION indicates that the RNA-trigger has detected a good pionic atom event.

### 3.5 VMEbus INTERFACE:

The VMEbus INTERFACE is responsible for the bi-directional communication between the $M-C A R D$ and a VMEbus-based computer. The VMEbus INTERFACE is located between the VMEbus and the 16bit on-board-bus of the M-CARD.

The M-CARD and the NN-CARDs are so called A16/D16 slave VME-CARDs (see the book THE VMEbus SPECIFICATION; ANSI/IEEE 1014-198 ; IEC821 \& IEC297). The A16/D16 means that the VME-CARD decodes only 16 address bit and uses only 16 data bit. The slave means that the VME-CARD cannot act as a bus master. The VME-CARD is designed for a short 16bit access, as proposed in the book THE VMEbus SPECIFICATION for I/O-cards. The access mode is coded in the Address Modifier (AM; 5bit). The VME-CARD listens only to short 16bit accesses, which have the $A M$-codes $2 D h$ or 29 . The data-width of the access is determined by the control lines DSO, DS1 and LWORD. The VME-CARD responds with a Data Transfer Acknowledge (DTACK-signal) only if a double-byte (16bit data) access is carried out. If a triple-byte, a quad-byte or a unaligned double-byte access is detected, the VME-CARD responds with a Bus Error (BERR-signal). The DATCK- and BERR-signals are indicated with a green and a red LED on the front panel.


Figure 12 The VME-address decoding scheme of the M-CARD: The M-Mode (W) and M-Status (R) register are located at address zero.

Figure 12 shows the VME-address decoding performed in the VMEbus Decoder \& Control Logic. Because the VME-CARD uses the double-byte (16bit data) access the VME-address-lines start from A1. The VME Card Number is set by the Hex Switch on the front panel of the MCARD. The VME Card Number can be set from Oh (Odec) to Fh (15dec), so sixteen different VME-CARDs can be addressed. Only if the VME Card Number matches the addresses A6...A9, and the addresses A10...A15 are zero, the M-CARD is selected (CARD_SELECTED). At the address zero of the $M$-CARD, the $M$-Mode and $M$-Status registers are located. The addresses $A 4$ and A5 determines which of the four Communication Chips (TU, PU, DU, CU) is selected. The lowest three addresses (A1...A3) are used to access the different 16 bit-registers of the selected Communication Chip.


Figure 13 VMEbus INTERFACE block diagram: The blocks marked with * are not part of the CPLD. The VMEbus INTERFACE performs the communication between the M-CARD and a VMEbus-based computer.

The block diagram of the VMEbus INTERFACE is shown in Figure 13. In the VMEbus TRANCEIVER the VMEbus TTL-signals are buffered via special transmitter- and receiverchips, as specified in the book THE VMEbus SPECIFICATION. To reduce the number of pins needed for the VME-CHIP the or-function of the address lines A10..15 is external implemented.

The M-Mode Register (write only) is used to select the different modes of the M-CARD:

- LOADED mode: The look up tables and the registers of the M-CARD are loaded and the $M$-CARD is ready to switch to the WORK mode. For safety reasons, writing to any register (except the NEXT/RESET in the TU) is suppressed in this mode. To switch to any TEST mode or to the LOAD mode is inhibited until the LOADED mode is cleared. The LOADED mode cannot be reached if any TEST mode or to the LOAD mode is active. These restrictions are implemented in the LOADED/WORK Safety Logic. The LOADED mode is displayed with a green $L E D$ on the frontpanel.
- WORK mode: To reach the WORK mode the LOADED mode has to be set previously. In this mode the M-CARD is ready to process events on the rising edge of the START-signal. The WORK-signal is wired to all of the four TU-, PU-, DU and CU-Chips. The WORK mode is displayed with a green $L E D$ on the frontpanel.
- SINGLE_SHOT mode: In this mode single events under real-time conditions can be analyzed via the VMEbus. The SINGLE_SHOT-signal is wired to the TU-Chip only. The SINGLE_SHOT mode is displayed with a yellow LED on the frontpanel.
- LOAD_SRAM mode: This mode allows the loading of the look up tables. The $L O A D \_S R A M$-signal is wired to the PU- and CU-Chips. The LOAD_SRAM mode is displayed with a red $L E D$ on the frontpanel.
- TEST_OVRALL mode: This mode allows the overall testing of the RNA trigger. The TEST_OVERALL-signal is wired to the PU-, DU and CU-Chips.
- TEST_TU mode: This mode enables the testing of the Timing-Unit. The TEST_TU-signal is wired to the TU-Chip only.
- TEST_PU mode: This mode allows the readout of the look up tables in the Pattern-Unit. The TEST_PU-signal is wired to the PU-Chip only.
- TEST_CU mode: This mode allows the readout of the look up tables in the C-CARDsUnit. The TEST_CU-signal is wired to the CU-Chip only.
- TEST_NN mode: This mode allows the testing of the NN-CARDs via the Pattern-Unit. The TEST_NN-signal is wired to the PU-Chip only.
- TEST_DU mode: This mode allows the testing of the Decision Logic as well as the output driver. The TEST_DU-signal is wired to the DU-Chip only.

If one of the TEST modes is selected the yellow LED TEST on the frontpanel is activated. With the restrictions mentioned above, the LOADED mode can be activated or deactivated via the LOADED_DEM bit in the M-Mode Register. A RESET-signal (active low) deactivates the LOADED mode immediately and a SYSFAIL-signal is generated on the VMEbus.

The RESET Logic $\&+5 V$ Control is responsible for the correct timing of the RESETsignal, for supervising the +5 V supply-voltage and for the power-on RESET. The RESET-signal is distributed all over the $M-C A R D$ and it sets the registers in a well defined state. There are three different sources, which can activate the RESET-signal:

- If the +5 V supply-voltage ( Vcc ) drops below a certain voltage (typical +4 V ).
- If a system-reset (/SYS_RESET) is received via the VMEbus.
- If the RESET_DEM bit of the M-Mode Register is set.

The $M$-Status Register (read only) allows to read the status of the $M$-CARD and the status of the SYSFAIL-signal on the VMEbus. The SYSFAIL-signal is active (low), if the M-CARD or one of the four $N N-C A R D s$ is no longer in the LOADED mode. The SYSFAIL-signal can easily be checked via the VMEbus for supervising the RNA-trigger.

The VMEbus INTERFACE is realized with TTL and HCT logic and a CPLD (Complex Programmable Logic Device). For further details see the CPLD documentation VME-CHIP.

## 4. NN-Patterns:

The $N N-C A R D$ s have a 55 bit wide $T T L$ pattern-input. The RNA-trigger application needs only a 26bit NN-PATTERN and the unused inputs are pulled to GND. The 26bit NN-PATTERN consists of the two SciFi-positions SciFi_POS1 (8bit) and SciFi_POS2 (8bit), one VL-position VL_1P or $V L \_2 P$ (5bit) and one VR-position VR_1P or VR_2P (5bit). The NN-PATTERN bits are terminated on the NN-CARDs with $180 \Omega$ to +5 V and $330 \Omega$ to GND. The SciFi_POS1/2 signals (16bit, NN PAT[0..15]) are fed in parallel to all of the four NN-CARDs; therefore only the last $N N-C A R D$ (number 4, most right) is terminated at these input bits (NN_PAT[0..15]). The VL/VR_1P/2P are driven for each $N N-C A R D$ separately by the $M-C A R D$ and therefore on all four $N N-C A R D s$ these input bits are terminated.

The bit allocation in the NN-PATTERN is with the MSB at the lowest position and the LSB at the highest position (e.g. NN_PAT[0] -> MSB of SciFi_POS1 / NN_PAT[7] -> LSB of SciFi_POS1).

The following different $N N$-PATTERNs are applied to the four $N N$-CARDs via the $R N A$ Backplane (SP 827A):

NN-CARD \#1: (Most left; no terminations on NN_PAT[0..15])

| NN_PAT[7...0] | $->$ | SciFi_POS1 (8bit) |
| :--- | :--- | :--- |
| NN_PAT[15...8] | -> | SciFi_POS2 (8bit) |
| NN_PAT[20...16] | -> | VL_1P (5bit) |
| NN_PAT[25...21] | $->$ | VR_1P (5bit) |

NN-CARD \#2: (No terminations on NN_PAT[0..15])

| NN_PAT[7...0] | $->$ | SciFi_POS1 (8bit) |
| :--- | :--- | :--- |
| NN_PAT[15...8] | -> | SciFi_POS2 (8bit) |
| NN_PAT[20...16] | -> | VL_1P (5bit) |
| NN_PAT[25...21] | -> | VR_2P (5bit) |

NN-CARD \#3: (No terminations on NN_PAT[0..15])

```
NN_PAT[7...0] -> SciFi_POS1 (8bit)
NN_PAT[15...8] -> SciFi_POS2 (8bit)
NN_PAT[20...16] -> VL_2P (5bit)
NN_PAT[25...21] -> VR_1P (5bit)
```

NN-CARD \#4: (Most right; terminations on NN_PAT[0..15])

```
NN_PAT[7...0] -> SciFi_POS1 (8bit)
NN_PAT[15...8] -> SciFi_POS2 (8bit)
NN_PAT[20...16] -> VL_2P (5bit)
NN_PAT[25...21] -> VR_2P (5bit)
```


# Fanout/Delay/Register description 

The Fanout/Delay/Register Card receives 64 data signals and a gate in differential ECL level (dECL) and is part of the hardware to be used on the Revised Neural Atomic trigger of the DIRAC experiment at CERN.

## Description

The main blocks of the FDR are the Fanout block and the Delay and Register block. Since the data signals are present prior to the trigger, a delay on the data is mandatory and in the end of the trigger the delayed pattern is registered until the next trigger.

## FANOUT and DELAY Registers.

To receive all the 64 differential inputs from flat cables of $1,27 \mathrm{~mm}$ spacing and still use a single 9U front panel a double connector was chosen from the KEL series. Each connector receives 16 dECL inputs and drives 16 dECL outputs, the 17th pair is connected to ground.

The signals arriving to the module are first converted to TTL prior to be latched on the Altera and fanout again to dECL levels for the next step of the electronics.

In the Altera the signals are delayed via a pipeline register with a programmable number of steps of 5 ns and latched on the end of the trigger signal to be used by the Concentrator card. The programmable delay is selected via two BCD selectors placed on the front panel, one for the dozens and the second one for the units.

The registered signals are connected to the concentrator card via P2. Row B holds the first 32 signals and C the higher 32 data delayed and registered. On the a Row A a copy of the first 16 signals is buffered to independently. The remaining pins are connected to ground. Figure 1 shows a schematic description of the FDR.

Figure 1:




